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Hybrid Active-and-Passive Relaying model for SWIPT-enabled 6G-IoT Greencom Networks

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Abstract—In order to support a massive number of resource-constrained Internet of Things (IoT) devices and machine-type devices, it is crucial to design future beyond 5G/6G wireless networks in an energy-efficient manner while incorporating the network extension methodologies. To this end, this letter proposes a novel two-hop hybrid active-and-passive relaying scheme to facilitate simultaneous wireless information and power transfer (SWIPT) considering both the Time-Switching (TS) and Power-Splitting (PS) receiver architectures, while dynamically modelling the involved dual-hop time-period (TP) metric. An optimization problem is formulated for the joint optimization of throughput, harvested energy, and transmit power of a SWIPT-enabled system with the proposed hybrid scheme, and is solved using a non-linear optimizer. Our numerical results in terms of weighted utility function show the superior performance of proposed hybrid scheme over passive repeater-only and active relay-only schemes, while also depicting their individual performance benefits over the corresponding benchmark SWIPT systems with the fixed-TP.

Index Terms—5G and beyond/6G wireless networks, greencom, IoT, passive repeater, relaying systems, SWIPT.

I. INTRODUCTION

The world witnesses the introduction of a novel generation of wireless communications in approximately every ten years. With the current speedy deployment of hardware for the fifth-generation (5G) of mobile wireless systems in 2020, we are now at a juncture to anticipate what lies ahead for the constitution of the sixth generation (6G) mobile cellular systems [1]. The incoming crucial updates to the hardware equipment for 5G may require dense infrastructure deployment [2]. In future, it will become important to leverage the 5G architecture for internet-of-things (IoT) systems, along with the introduction of some methods to increase the coverage area as well as considering energy harvesting (EH) frameworks embracing the concept of Green Communications (Greencom)-IoT systems.

To enhance the network coverage of wireless communications systems, certain well-known methods such as passive repeater (PR) and/or relaying have shown immense potential [3]. A PR is a battery-less device used to reflect the transmit signal (with encoded data) towards the desired user, where it may be utilized to energize the IoT devices as well as to enable the deployment of battery-free devices in future networks [4]. Noticeably, backscattering may be considered as a special form of PR. In case of an active relay (AR), the transmitted signal from the source is received at the relay device and re-transmitted to the end-user after either a direct amplification according to the amplify-and-forward (AF) protocol [5], or a decoding or quantization process before its re-transmission as a re-encoded signal with message suppression, according to the decode-and-forward (DF) protocol [6]. Regarding the

facilitation of EH at the end-user, the simultaneous wireless information and power transfer (SWIPT) technique seems promising [7]. In this regard, the power-splitting (PS) and time-switching (TS) schemes have been widely considered to deploy SWIPT receivers in practice.

Recent developments in the area of PR and AR type of scheme have focused on important aspects such as cooperation enhancement, however, with the consideration of a linear EH model [8]. In [9], the authors presented an interesting framework for wireless energy beamforming in the reflector-and-relay based communications system, wherein a non-linear EH mechanism is considered at the intermediary device but not at the end-user. Additionally, most related works focus on the maximization of either the signal-to-noise ratio (SNR) [10], or the throughput [11]. However, it is important to optimize the joint metrics of data throughput, harvested energy and transmit power of the SWIPT-enabled 6G systems. On one hand, it is noteworthy that the reflector-only mechanism may not be suitable for dense deployment scenarios [4], [12]. Whereas on the other hand, the power-dependence of relays is critical in the relaying-only scenarios, wherein a power-failure may lead to the collapse of the entire wireless system without any backup. In this regard, the hybrid PR and AR based scheme may ensure a sustainable wireless ecosystem where both devices may supplement or complement each other accordingly. Moreover, in the context of the hybrid framework, a comparative study between the PS and TS SWIPT schemes with non-linear EH model along with the design of dynamic dual-hop time-period (TP) has not yet been investigated in the literature so far.

In this letter, we present a dual-hop system wherein a transmit source transfers both information and energy to a SWIPT-enabled end-user employing either PS or TS architecture. We investigate three distinct Greencom systems incorporating a PR node, an AR node, and a novel hybrid active-and-passive relaying node, respectively; which either reflects and/or re-transmits the information-and-energy signal to the end-user, in addition to the weak direct link. The novelty and main contributions of this work are listed as follows.

(i) In contrast to the existing reflector-only and relaying-only techniques, we propose a novel hybrid active-and-passive relaying scheme to facilitate SWIPT to a PS- or TS-enabled end-user, along with the dynamic designing of dual-hop TP, under certain receive processing assumptions.

(ii) We formulate a novel problem (incorporating the three systems) to maximize the weighted utility function comprising data throughput, harvested energy and transmit power, subjected to some quality-of-service (QoS) constraints. Unlike the other works that assume equal time-intervals in the two hops (also considered herein as the benchmarks), we present a framework to dynamically design the TP for the dual-hop link, along with the computation of the PS or TS factors.

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(iii) The effectiveness of the proposed hybrid active-and-passive relaying scheme is shown over the reflector-only and relay-only schemes for both PS and TS SWIPT schemes via numerical analysis, with individual benefits shown over their respective benchmark designs having a fixed TP.

The remainder of this paper is organized as follows. Section II presents a description of the system model and Section III provides an analysis of the considered Greencom networks. Section IV focuses on the joint optimization of data throughput, harvested energy and transmit power. Section V presents the simulation results. Finally, Section VI concludes the paper.

II. SYSTEM MODEL

We consider three wireless Greencom network scenarios for investigation, with a single transmit source, a single reflection and/or re-transmission device (i.e., either PR or AR or both), and a single user at the receiving end, as represented in Fig. 1. The transmit source communicates with the end-user via two half-duplex communication links within a single time-slot. In particular, besides the conventional direct link, we assume the availability of a couple of intermediary links to assist in the delivery of the desired transmit signal. The end-user is assumed to be capable of performing both information decoding (ID) and energy harvesting (EH) simultaneously.

Let us define P_T as the transmit power at the source, B as the overall bandwidth, and T as the TP for further analysis. In each system, the direct link is assumed to be active throughout the TP, with a pre-defined channel coefficient of $h_{1,D}$.¹ The transmit source emits the signal intended for the corresponding SWIPT-enabled end-user. The process is carried out using a direct link through the whole TP, with additional dual-hop assistance via a reflecting and/or re-transmitting device over the same TP. More specifically, the source transmits a symbol $s \in \mathbb{C}$, which is received by the destination and the reflecting and/or re-transmitting device. The propagation path of the signal is sufficiently small and the intermediary device does not introduce any significant delays. The destination node is considered to utilize delay compensation methods to synchronize and coherently process the received signals from the second hop and the direct link. Without loss of generality, we assume $\mathbb{E}[|s|^2] = 1$, with operation $\mathbb{E}[\cdot]$ denoting the expectation value. The signal received by the end-user via direct link is defined as: $y_{1,D} = \sqrt{P_T} h_{1,D} s + n_{u_a}$, where n_{u_a} is the additive white Gaussian noise (AWGN) at the end-user which is an independent and identically distributed (i.i.d.) complex Gaussian random variable with zero mean and variance $\sigma_{n_{u_a}}^2$.

Concerning the SWIPT-enabled receiver device, two different SWIPT schemes namely PS or TS, respectively, may be adopted. For the PS scheme, an optimal fraction of the received signal is provided to the information decoder and the remaining part is provided to the energy harvester. In this case, the PS ratio is denoted by ρ , where $0 \leq \rho \leq 1$. Alternatively, when the receiver incorporates the TS scheme, we define a time switching ratio, τ , where $0 \leq \tau \leq 1$. In particular, for the first fraction of time period, all the received signal power is used for harvesting energy, whereas during the remaining time, information decoding from the received signal takes place.

¹For analytical convenience, we assume that the channel state information (CSI) is known to the intended devices.

Fig. 1: Dual hop system with a PR and/or AR device.

In order to estimate the amount of harvested energy at the end-user, we adopt the sigmoidal function based non-linear EH model [13], [14], defined as

$$\mathcal{E}(x) = \frac{\mathcal{E}'}{1-\phi} \cdot \left(\frac{1}{1+e^{(-\alpha x + \alpha\beta)}} - \phi \right), \quad (1)$$

where x is the input power at the energy harvesting module, $\phi \triangleq \frac{1}{1+\exp(\alpha\beta)}$, the constant $\mathcal{E}'$ is obtained by determining the maximum harvested energy on the saturation of the energy harvesting circuit, and α and β are specific to the capacitor and diode turn-on voltage metrics at the EH circuit. The harvested energy using the PS and TS schemes are respectively given by

$$\mathcal{E}^{\text{PS}}(x, \rho) = \frac{\mathcal{E}'}{1-\phi} \cdot \left(\frac{1}{1+e^{(-\alpha\rho x + \alpha\beta)}} - \phi \right), \quad (2)$$

$$\mathcal{E}^{\text{TS}}(x, \tau) = \tau \cdot \frac{\mathcal{E}'}{1-\phi} \cdot \left(\frac{1}{1+e^{(-\alpha x + \alpha\beta)}} - \phi \right). \quad (3)$$

The energy harvested at the end-user using the PS and TS schemes via direct link signal is given by $\mathcal{E}_{1,D}^{\text{PS}}(\mathbb{E}[|y_{1,D}|^2], \rho)$ and $\mathcal{E}_{1,D}^{\text{TS}}(\mathbb{E}[|y_{1,D}|^2], \tau)$, computed according to (2) and (3), respectively. Further, we assume that the normalized time slots use the terms *power* and *energy* interchangeably.

The effective signal-to-noise ratio (SNR) via direct link as seen at the ID branch of the end-user, respectively for PS and TS schemes, is given by

$$\Upsilon_{1,D}^{\text{PS}} = \frac{(1-\rho) P_T |h_{1,D}|^2}{(1-\rho) \sigma_{n_{u_a}}^2 + \sigma_{n_{u_{ID}}}^2}, \text{ and } \Upsilon_{1,D}^{\text{TS}} = \frac{P_T |h_{1,D}|^2}{\sigma_{n_{u_a}}^2 + \sigma_{n_{u_{ID}}}^2}, \quad (4)$$

where $n_{u_{ID}} \in \mathbb{CN}(0, \sigma_{n_{u_{ID}}}^2)$ is the noise introduced by the baseband processing circuit. The effective throughput achieved at the end-user, combining the PS and TS schemes is given by

$$R_{1,D} = \begin{cases} R_{1,D}^{\text{PS}} = B \log_2(1 + \Upsilon_{1,D}^{\text{PS}}) \\ R_{1,D}^{\text{TS}} = (1-\tau) B \log_2(1 + \Upsilon_{1,D}^{\text{TS}}) \end{cases} \quad (5)$$

III. ANALYSIS OF GREENCOM NETWORK SCENARIOS

In this section, we present the discussion on the considered Greencom network scenarios in detail. To proceed, we define the dynamic time-splitting (DTS) ratio as δ , where $0 < \delta \leq 1$, to assist the dual-hop mechanism. Specifically, we assume that the first-hop spans over δT of the TP while in the second hop, the remainder of the TP, i.e., $(1-\delta)T$, is allocated for successful reflection and/or re-transmission of the intended signal from the PR and/or AR device(s) to the end-user.

A. Traditional passive repeater or active relay-based systems

In this case, we assume the availability of either reflection-antenna equipped PR device as the reflector node, or a single antenna based AR node as a re-transmission device, on the second hop. Further, we refer the two above-mentioned systems as PR device-based Greencom network

(PGN) and AR device-based Greencom network (AGN), respectively. The channel coefficient in the first phase is defined as h_1 . We represent the signal seen at the reflecting/re-transmitting device as: $r_1 = \sqrt{P_T} h_1 s + n_1$, where n_1 is the AWGN introduced due to the reflector/relay, which is an i.i.d. complex Gaussian random variable with zero mean and variance $\sigma_{n_1}^2$. In this context, we express $(r_1, h_1, n_1) := \{(r_{1,P}, h_{1,P}, n_{1,P}) : \text{for PGN}; (r_{1,R}, h_{1,R}, n_{1,R}) : \text{for AGN}\}$.

The SNR and the effective data throughput expressions are respectively defined as

$$\Upsilon_1 = \frac{P_T |h_1|^2}{\sigma_{n_1}^2}, \text{ and } R_1 = \delta \mathcal{B} \log_2(1 + \Upsilon_1), \quad (6)$$

where $(\Upsilon_1, R_1, \sigma_{n_1}^2) := \{(\Upsilon_{1,P}, R_{1,P}, \sigma_{n_{1,P}}^2) : \text{for PGN}; (\Upsilon_{1,R}, R_{1,R}, \sigma_{n_{1,R}}^2) : \text{for AGN}\}$.

In the second hop, the channel coefficient is defined as h_2 . In case of PGN, the intended signal from the PR is reflected to the end-user with the corresponding efficiency coefficient of $\sqrt{\eta}$, where $0 < \sqrt{\eta} \leq 1$. Whereas in the case of AGN, the AR re-transmits the signal after scaling it by a complex amplification coefficient w . In order to ensure feasibility of the AGN system, we impose an upper bound on the total relay power defined by: $0 < |w|^2 \leq P_R$, where $P_R = \frac{P^* - P_T}{P_T |h_{1,R}|^2 + \sigma_{n_{1,R}}^2}$ is the maximum overall available power at the relay, with the transmitter-relay system bounded by an overall power of P^* , such that $P^* > \max(P_T, P_R)$. To proceed, we define the following metric:

$$\Xi = \begin{cases} \sqrt{\eta} : \text{PR's Reflection Coefficient in PGN} \\ w : \text{AR's Amplification Coefficient in AGN} \end{cases} \quad (7)$$

The received signal as seen at the end-user via reflection or re-transmission from the PR or AR device, respectively, is represented as $y_2 = \Xi h_2 r_1 + n_{u_a}$, where $(y_2, h_2, r_1) := \{(y_{2,P}, h_{2,P}, r_{1,P}) : \text{for PGN}; (y_{2,R}, h_{2,R}, r_{1,R}) : \text{for AGN}\}$.

We define the SNR obtained at the ID block of the end-user via second hop link, according to the PS and TS schemes, as

$$\Upsilon_2^{\text{PS}} = \frac{|\Xi|^2 (1-\rho) P_T |h_1|^2 |h_2|^2}{(1-\rho) (|\Xi|^2 |h_1|^2 \sigma_{n_1}^2 + \sigma_{n_{u_a}}^2) + \sigma_{n_{ID}}^2}, \quad (8)$$

$$\Upsilon_2^{\text{TS}} = \frac{|\Xi|^2 P_T |h_1|^2 |h_2|^2}{|\Xi|^2 |h_1|^2 \sigma_{n_1}^2 + \sigma_{n_{u_a}}^2 + \sigma_{n_{ID}}^2}, \quad (9)$$

where $(\Upsilon_2^{\text{PS}}, \Upsilon_2^{\text{TS}}, h_1, h_2, \sigma_{n_1}^2) := \{(\Upsilon_{2,P}^{\text{PS}}, \Upsilon_{2,P}^{\text{TS}}, h_{1,P}, h_{2,P}, \sigma_{n_{1,P}}^2) : \text{for PGN}; (\Upsilon_{2,R}^{\text{PS}}, \Upsilon_{2,R}^{\text{TS}}, h_{1,R}, h_{2,R}, \sigma_{n_{1,R}}^2) : \text{for AGN}\}$. The effective throughput achieved at the end-user, incorporating the respective PS and TS schemes is defined as

$$R_2 = \begin{cases} R_2^{\text{PS}} = (1-\delta) \mathcal{B} \log_2(1 + \Upsilon_2^{\text{PS}}) \\ R_2^{\text{TS}} = (1-\tau)(1-\delta) \mathcal{B} \log_2(1 + \Upsilon_2^{\text{TS}}) \end{cases}, \quad (10)$$

where $(R_2, R_2^{\text{PS}}, \Upsilon_2^{\text{PS}}, R_2^{\text{TS}}, \Upsilon_2^{\text{TS}}) := \{(R_{2,P}, R_{2,P}^{\text{PS}}, \Upsilon_{2,P}^{\text{PS}}, R_{2,P}^{\text{TS}}, \Upsilon_{2,P}^{\text{TS}}) : \text{for PGN}; (R_{2,R}, R_{2,R}^{\text{PS}}, \Upsilon_{2,R}^{\text{PS}}, R_{2,R}^{\text{TS}}, \Upsilon_{2,R}^{\text{TS}}) : \text{for AGN}\}$.

The PS and TS expressions for harvested energy are denoted as $\mathcal{E}_2^{\text{PS}}(\mathbb{E}[|y_2|^2], \rho)$ and $\mathcal{E}_2^{\text{TS}}(\mathbb{E}[|y_2|^2], \tau)$, computed according to (2) and (3), respectively, where $(\mathcal{E}_2^{\text{PS}}, \mathcal{E}_2^{\text{TS}}, y_2) := \{(\mathcal{E}_{2,P}^{\text{PS}}, \mathcal{E}_{2,P}^{\text{TS}}, y_{2,P}) : \text{for PGN}; (\mathcal{E}_{2,R}^{\text{PS}}, \mathcal{E}_{2,R}^{\text{TS}}, y_{2,R}) : \text{for AGN}\}$.

At the end-user, the overall data throughput achieved incorporating the yield from the second hop and direct link, for respective PS and TS schemes, is defined as

$$R_U = \begin{cases} R_U^{\text{PS}} = R_{1,D}^{\text{PS}} + R_2^{\text{PS}} \\ R_U^{\text{TS}} = R_{1,D}^{\text{TS}} + R_2^{\text{TS}} \end{cases}, \quad (11)$$

where $(R_U, R_U^{\text{PS}}, R_U^{\text{TS}}) := \{(R_{U,P}, R_{U,P}^{\text{PS}}, R_{U,P}^{\text{TS}}) : \text{for PGN}; (R_{U,R}, R_{U,R}^{\text{PS}}, R_{U,R}^{\text{TS}}) : \text{for AGN}\}$. The effective harvested en-

ergy expression incorporating both PS and TS schemes is:

$$E_U = \begin{cases} E_U^{\text{PS}} = \mathcal{E}_{1,D}^{\text{PS}}(\mathbb{E}[|y_{1,D}|^2], \rho) + (1-\delta) \mathcal{E}_2^{\text{PS}}(\mathbb{E}[|y_2|^2], \rho) \\ E_U^{\text{TS}} = \mathcal{E}_{1,D}^{\text{TS}}(\mathbb{E}[|y_{1,D}|^2], \tau) + (1-\delta) \mathcal{E}_2^{\text{TS}}(\mathbb{E}[|y_2|^2], \tau) \end{cases} \quad (12)$$

where $(E_U, E_U^{\text{PS}}, E_U^{\text{TS}}) := \{(E_{U,P}, E_{U,P}^{\text{PS}}, E_{U,P}^{\text{TS}}) : \text{for PGN}; (E_{U,R}, E_{U,R}^{\text{PS}}, E_{U,R}^{\text{TS}}) : \text{for AGN}\}$.

B. Proposed Hybrid Active-and-Passive Relaying Scheme

We present herein the proposed hybrid active-and-passive (HAP) relaying scheme in Greencom networks to facilitate SWIPT to the end-user over the second hop. In this case, the HAP relaying node receives the transmit signal on both the PR and AR system equipped with single antenna each, over the above-defined channel coefficients of $h_{1,P}$ and $h_{1,R}$, respectively. The DTS ratio definition is same as defined before. The signals, SNR and throughput expressions seen at the respective PR and the AR nodes are similar to the ones defined in the previous sections.²

Over the second hop, the PR device uses the coefficient $\sqrt{\eta}$ to reflect the main transmit signal while the AR employs the AF-protocol to boost the transmit signal with the help of w and re-transmit the same to the end-user over the second hop. For the remainder of TP, i.e., $(1-\delta)T$, the re-directed signals from the PR and AR experiences channel coefficients of $h_{2,P}$ and $h_{2,R}$, respectively, with an assumption of negligibly small delay which can be ignored for the further analysis. The corresponding signals, SNR and throughput expressions are similar as defined before. The overall effective throughput and the harvested energy expressions achieved at the end-user, incorporating the PS and TS schemes are respectively given by

$$R_{U,PR} = \begin{cases} R_{U,PR}^{\text{PS}} = R_{1,D}^{\text{PS}} + R_{2,P}^{\text{PS}} + R_{2,R}^{\text{PS}} \\ R_{U,PR}^{\text{TS}} = R_{1,D}^{\text{TS}} + R_{2,P}^{\text{TS}} + R_{2,R}^{\text{TS}} \end{cases},$$

$$E_{U,PR} = \begin{cases} E_{U,PR}^{\text{PS}} = \mathcal{E}_{1,D}^{\text{PS}}(\mathbb{E}[|y_{1,D}|^2], \rho) + (1-\delta) \mathcal{E}_{2,P}^{\text{PS}}(\mathbb{E}[|y_{2,P}|^2], \rho) \\ \quad + (1-\delta) \mathcal{E}_{2,R}^{\text{PS}}(\mathbb{E}[|y_{2,R}|^2], \rho) \\ E_{U,PR}^{\text{TS}} = \mathcal{E}_{1,D}^{\text{TS}}(\mathbb{E}[|y_{1,D}|^2], \tau) + (1-\delta) \mathcal{E}_{2,P}^{\text{TS}}(\mathbb{E}[|y_{2,P}|^2], \tau) \\ \quad + (1-\delta) \mathcal{E}_{2,R}^{\text{TS}}(\mathbb{E}[|y_{2,R}|^2], \tau) \end{cases}.$$

IV. PROBLEM FORMULATION AND SOLUTION

Herein, we present the problem to jointly optimize the data throughput, harvested energy and transmit power, subjected to QoS constraints for each considered schemes (viz., PGN, AGN, and HAP). Before proceeding to the main hypothesis, we introduce some variables to simplify the problem representation.

A. Variable definitions to assist the problem formulation

In this section, we define some additional parameters corresponding to the overall transmit power, data throughput and harvested energy expressions obtained at the end-user, the DTS ratios, the relay amplification factor and the TS and PS ratios; to refer to the aforementioned frameworks of PGN, AGN, and HAP altogether. In this context, the variables are sequentially defined in Table I.

We additionally consider another parameter to assist the PS and TS ratio metrics, defined as follows.

$$\theta := \{\rho : \text{for PS Scheme}; \tau : \text{for TS Scheme}\}. \quad (13)$$

²Note that the PR and AR devices in the proposed HAP scheme may be co-located or not. However, for analytic convenience related to the DTS design, we assume a co-located hybrid PR and AR set-up.

Parameter System	P	R_I	R_{II}	R_U	E_U
PGN	P_T	$R_{1,P}$	$R_{2,P}$	$R_{U,P}$	$E_{U,P}$
AGN	$P_T + P_R$	$R_{1,R}$	$R_{2,R}$	$R_{U,R}$	$E_{U,R}$
HAP	$P_T + P_R$	$R_{1,P}$ $R_{1,R}$	$R_{2,P}$ $R_{2,R}$	$R_{U,P,R}$	$E_{U,P,R}$

TABLE I: Parameter definitions to assist problem formulation.

It is clear that the newly defined parameters will adopt values according to the type of scenario chosen. Thus, we mathematically formulate the optimization problem in the following while incorporating the parameters defined in Table I and (13).

B. Optimization problem with weighted utility function

We formulate the problem with weighted utility function to jointly optimize the data throughput, harvested energy and total transmit power of the source, while ensuring that the demanded throughput and the harvested power at the destination node are both individually above the given thresholds. The overall optimization problem incorporating the PGN, AGN and HAP systems is analytically represented as

$$(P1) : \max_{P_T, \delta, \theta, w} \frac{\omega_R R_U + \omega_E E_U}{\omega_P P_T} \quad (14)$$

$$\text{subject to : } (C1) : R_I \geq R_{II}, \quad (C2) : R_U \geq \varsigma, \quad (15)$$

$$(C3) : E_U \geq \xi, \quad (C4) : 0 \leq \delta \leq 1, \quad (16)$$

$$(C5) : 0 \leq \theta \leq 1, \quad (C6) : 0 < P \leq P^*, \quad (17)$$

$$(C7)^* : 0 \leq |w|^2 \leq P_R, \quad (18)$$

where ς and ξ are respectively the minimum demanded throughput and harvested energy by the destination SWIPT-capable user, ω_R , ω_E and ω_P are the constant weighing factors corresponding to the data throughput, harvested energy and the transmit power, respectively. We use θ to interchangeably refer to the PS or TS splitting factor ρ or τ , respectively. In order to ensure feasibility during the second hop transmission, we make use of (C1). The constraints (C2) and (C3) are the minimum thresholds on the overall data rate and overall harvested energy, respectively. The upper and lower bounds of the DTS fraction and the SWIPT splitting factors are represented in (C4) and (C5) respectively. The constraint with maximum transmit power budget of the system is incorporated using (C6). Note that the variable w and (C7)* are considered only during the analysis of AGN and HAP systems. In the following, we analyse (P1) and seek a suitable solution.

C. Solution to the above-mentioned problem

The problem (P1) is difficult to solve, since it has a non-convex form for both TS and PS schemes. In order to obtain an analytical solution for the joint objective optimization problem in (P1), we assume $\mathcal{D}$ as the set of all possibilities of P_T, δ, θ and w which satisfy the constraints (C4) – (C7). Then the corresponding Lagrange dual function is written as

$$J(\mathbf{\Lambda}) \triangleq \max_{\{P_T, \delta, \theta, w\} \in \mathcal{D}} \mathcal{L}(P_T, \delta, \theta, w; \mathbf{\Lambda}), \quad (19)$$

where the Lagrangian is expressed as

$$\begin{aligned} \mathcal{L}(P_T, \delta, \theta, w; \mathbf{\Lambda}) = & F(P_T, \delta, \theta, w) - \lambda_1 G(P_T, \delta, \theta, w) \\ & - \lambda_2 H(P_T, \delta, \theta, w) - \lambda_3 I(P_T, \delta, \theta, w), \end{aligned} \quad (20)$$

with $\mathbf{\Lambda} = (\lambda_1, \lambda_2, \lambda_3) \succeq 0$ denoting the vector of the dual variables associated with the constraints (C1) – (C3),

$F(P_T, \delta, \theta, w)$	$G(P_T, \delta, \theta, w)$	$H(P_T, \delta, \theta, w)$	$I(P_T, \delta, \theta, w)$
$\frac{\omega_R R_U + \omega_E E_U}{\omega_P P_T}$	$R_{II} - R_I \leq 0$	$\varsigma - R_U \leq 0$	$\xi - E_U \leq 0$

TABLE II: Definitions to comprise the Lagrangian expression.

respectively, while the related functions are as defined in Table II. The dual optimization problem is hence represented as

$$(P2) : \min_{\mathbf{\Lambda}} J(\mathbf{\Lambda}) \quad (21)$$

$$\text{subject to : } \mathbf{\Lambda} \succeq 0. \quad (22)$$

Since it is explicit that a dual problem is always convex by definition [15], therefore the gradient or subgradient-based methods can be used to minimize $J(\mathbf{\Lambda})$ with guaranteed convergence. In this regard, an alternating optimization of the involved parameters is performed in an iterative manner, until the convergence of the weighted utility function value. The intermediary closed-form expressions for the involved parameters may be obtained using the Karush-Kuhn-Tucker (KKT) conditions [15]. Discussion on the obtained numerical solutions is provided in the succeeding section.

V. SIMULATION RESULTS

In this section, we present a comparative study among the HAP, AGN and PGN systems. The three cases are analyzed using the MATLAB R2019b, with optimization performed via *fmincon*(.) solver present in the optimization toolbox [16], where the solutions are obtained with the help of interior point algorithm [15]. We assume an ITU-R outdoor framework (site-general model for propagation within street canyons) [17] to generate channel realizations with the path-loss exponent:

$$PL(d, f) = 10a \log_{10}(d) + b + 10c \log_{10}(f) + N(0, \sigma) \text{ dB}, \quad (23)$$

where d is distance between the transmitting and receiving stations (m), f is the operating frequency (GHz), the coefficients a , b , and c are associated with the increase of the path loss with distance, the offset value of the path loss, and the increase of the path loss with frequency, respectively, and $N(0, \sigma)$ is a zero mean Gaussian random variable with a standard deviation σ (dB). The channels coefficients for the direct link and the dual-hop links are generated accordingly. Specifically, we choose: $f = 24$ GHz, d is randomly chosen between 8 m and 10 m in case of PR/AR device and randomly between 12 m and 15 m for the end-user with respect to the transmitter, $a = 2.12$, $b = 29.2$, $c = 2.11$ and $\sigma = 5.06$ dB. The simulation results presented in this section assume an overall bandwidth of $B = 50$ MHz. We set $\sigma_{n_B}^2 = \sigma_{n_R}^2 = \sigma_{n_{u_a}}^2 = -150$ dBW/Hz, and $\sigma_{n_{u_{I,D}}}^2 = -110$ dBW/Hz. The PR's reflection efficiency coefficient is chosen as $\eta = 0.78$ (corresponding to 1.1 dB loss [18]). The constants corresponding to the non-linear EH circuit are chosen as $\mathcal{E}' = 2.8$ mJ, $\alpha = 1500$, and $\beta = 0.0022$ [13], [14]. The harvested energy demand is assumed to be $\xi = -70$ dBW, and the weighing factors are chosen as $\omega_R = 1$ units/Gbit, $\omega_E = 1$ units/nJ, and $\omega_P = 1$ units/W. We use two test cases for the data demands of $\varsigma = 1.00$ Gbits and $\varsigma = 1.05$ Gbits. As a benchmark scheme, we assume equal time distribution for both the hops in all the cases, i.e., fixed $\delta = 0.5$. Furthermore, an average of 1000 random channel realizations is presented for each experiment.

In Fig. 2 and Fig. 3, we present the performance analyses of HAP, AGN and PGN in terms of the weighted utility function for the PS and TS schemes, respectively, versus the maximum limitation on the total transmit power. We observe that all the

Fig. 2: Performance analysis of weighted utility function for PS scheme versus the maximum limitation on the total transmit power.

Fig. 3: Performance analysis of weighted utility function for TS scheme versus the maximum limitation on the total transmit power.

considered schemes can provide appreciable gains over their respective benchmark methods. The weighted utility function values for both PS and TS increases with growing overall maximum transmit power budget (P^*) of the system. On the other hand, the performance is adversely affected when the data demand is increased from $\varsigma = 1.00$ Gbits to $\varsigma = 1.05$ Gbits, which is due to the joint optimization in the weighted utility function. Particularly, this effect is due to the well-known rate-energy trade-off incurred because of the joint maximization of data and harvested energy along with the minimization of total transmit power. In the case of PS, we notice improvements of 10% and 12.5% for HAP over the AGN and PGN scenarios respectively for the data demand of $\varsigma = 1.00$ Gbits, and improvements of 2% and 4% respectively for the data demand of $\varsigma = 1.05$ Gbits, with an average decrement of 53% for a transition from former to the latter demand. Concerning the TS, we observe gains of 25% and 28% for HAP over the AGN and PGN scenarios respectively for the data demand of $\varsigma = 1.00$ Gbits, and improvements of 11.5% and 12.5% respectively for the data demand of $\varsigma = 1.05$ Gbits, with an average overall

decrement of 80.5% for a transition from $\varsigma = 1.00$ Gbits to $\varsigma = 1.05$ Gbits. Clear benefits for HAP are observed over the AGN and PGN counterparts, with PS scheme outperforming the TS.

VI. CONCLUSIONS

In this letter, we investigated a hybrid active-and-passive (HAP) relaying scheme to not only improve the network coverage, but also to facilitate SWIPT in a 5G and beyond/6G Greencom environment, where the end-users are capable of decoding information and harvesting energy simultaneously, according to the PS and TS schemes while using the non-linear EH device for the latter. We formulated an optimization problem to jointly address the respective demands of data throughput, harvested energy and total transmit power, under some QoS constraints. We showed significant benefits of adopting HAP over the AGN and PGN counterpart systems with the help of numerical results, where the PS scheme was found to outperform the TS concerning SWIPT at the end-user.

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